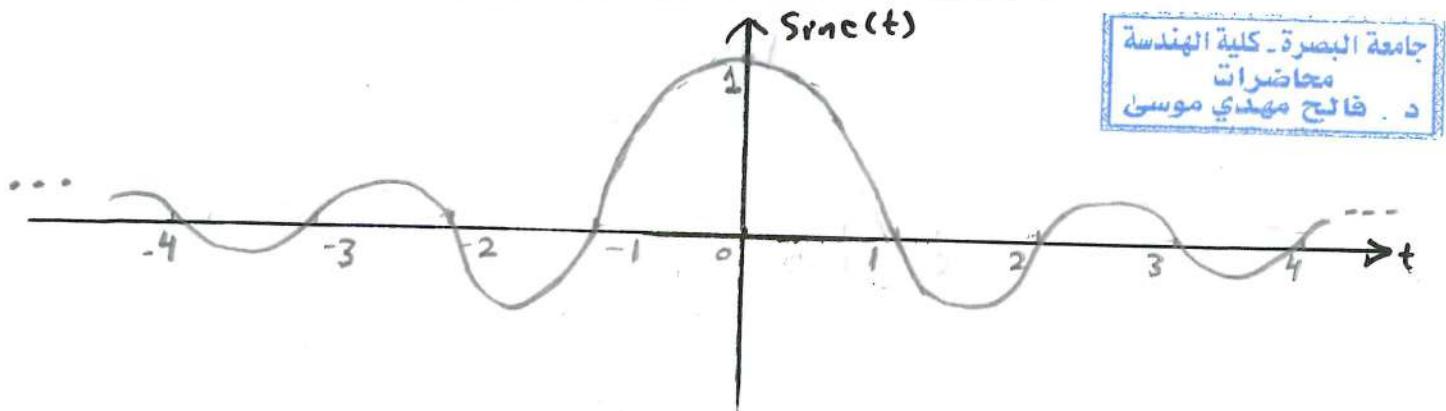


[d] Sinc Function :

It is very important signal especially in digital communication systems:

$$\text{Sinc}(t) = \frac{\sin(\pi t)}{\pi t}$$



جامعة البصرة - كلية الهندسة
محمادرات
د. فلاح مهدي موسى

- * It has even symmetry w.r.t (t).
- * Its max. value occurs at $t=0$.
- * It has zero values at $\pi t = \pm\pi, \pm 2\pi, \pm 3\pi, \dots$
 $t = \pm 1, \pm 2, \pm 3, \dots$

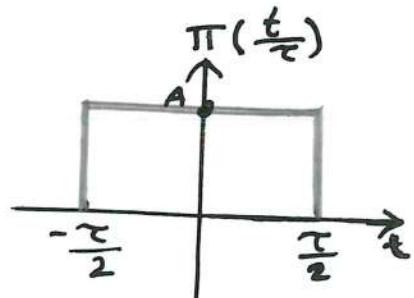
for example, the first zero of $\text{Sinc}(\frac{3}{7}\omega)$:

$$\text{Sinc}(\frac{3}{7}\omega) = \frac{\sin(\pi \frac{3}{7}\omega)}{\pi \frac{3}{7}\omega}$$

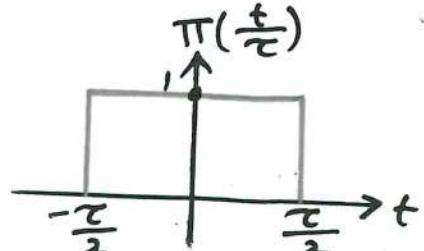
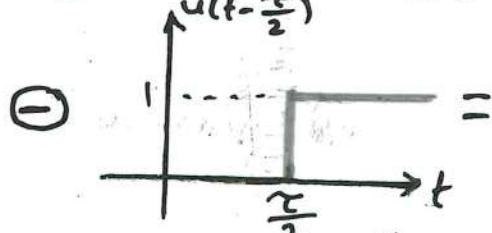
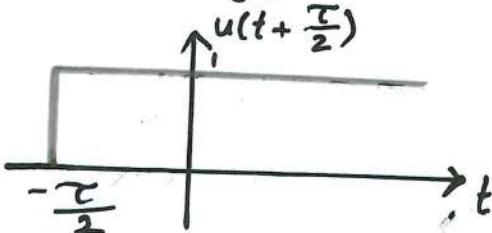
$$\sin(\pi \frac{3}{7}\omega) = 0 \quad \text{when} \quad \frac{\pi}{7} \frac{3}{7}\omega = \pi \\ \therefore \frac{3}{7}\omega = 1 \Rightarrow \boxed{\omega = \frac{7}{3}}$$

[e] Rectangular Pulse $\Pi(\frac{t}{\tau})$:

$$x(t) = A \Pi(\frac{t}{\tau}) = \begin{cases} A & |t| < \frac{\tau}{2} \\ 0 & |t| > \frac{\tau}{2} \end{cases}$$

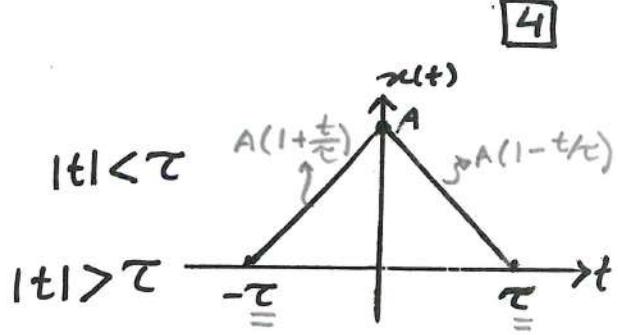


$$\Pi(\frac{t}{\tau}) = u(t + \frac{\tau}{2}) - u(t - \frac{\tau}{2})$$



Triangular Pulse $\Lambda(\frac{t}{\tau})$:

$$x(t) = A \Lambda\left(\frac{t}{\tau}\right) = \begin{cases} A(1 - \frac{|t|}{\tau}) & |t| < \tau \\ 0 & |t| > \tau \end{cases}$$

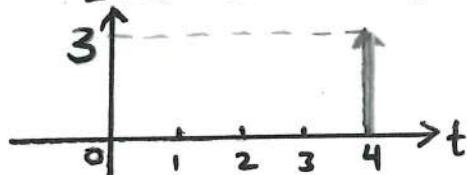


Example: Sketch the following signals:

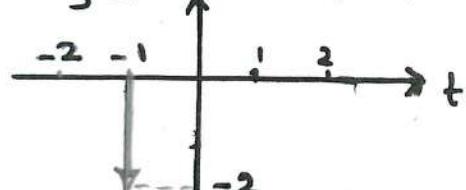
a) $x_1(t) = \delta(t)$



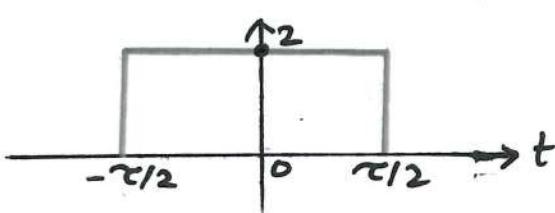
, $x_2(t) = 3\delta(t-4)$



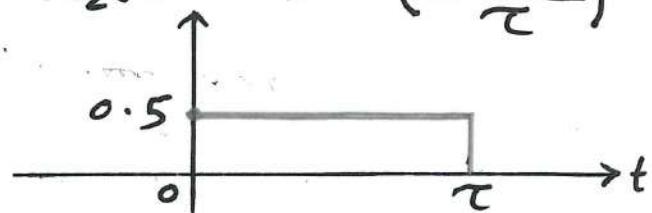
- $x_3(t) = -2\delta(t+1)$



b) $x_1(t) = 2\pi(\frac{t}{\pi/2})$



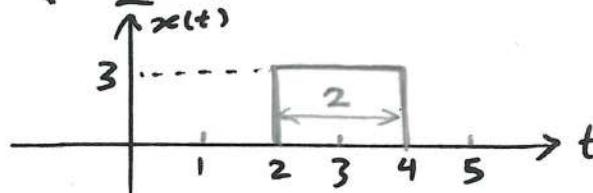
- $x_2(t) = 0.5\pi(\frac{t-\frac{\pi}{2}}{\pi})$



c) $x(t) = 3\pi(\frac{8t-24}{16})$

$$x(t) = 3\pi\left(\frac{8(t-\frac{24}{8})}{16}\right) = 3\pi\left(\frac{t-\frac{24}{8}}{\frac{16}{8}}\right)$$

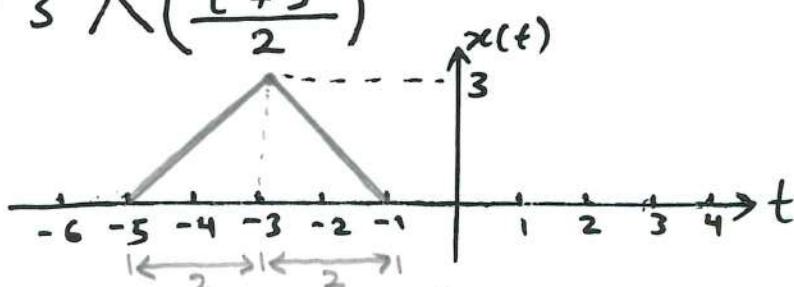
$$x(t) = 3\pi\left(\frac{t-3}{2}\right)$$



d) $x(t) = 3\Lambda(\frac{8t+24}{16})$

$$x(t) = 3\Lambda\left(\frac{8(t+\frac{24}{8})}{16}\right) = 3\Lambda\left(\frac{t+\frac{24}{8}}{\frac{16}{8}}\right)$$

$$x(t) = 3\Lambda\left(\frac{t+3}{2}\right)$$



2. Phasor and Spectrum:

A Single-Sided Spectrum:

Consider the conventional sinusoidal signal waveform:

$$x(t) = A \cos(\omega_0 t + \phi)$$

$$\text{OR } x(t) = A \cos(2\pi f_0 t + \phi)$$

where A is the amplitude

ϕ is the phase angle

ω_0 is the angular frequency (rad./sec.)

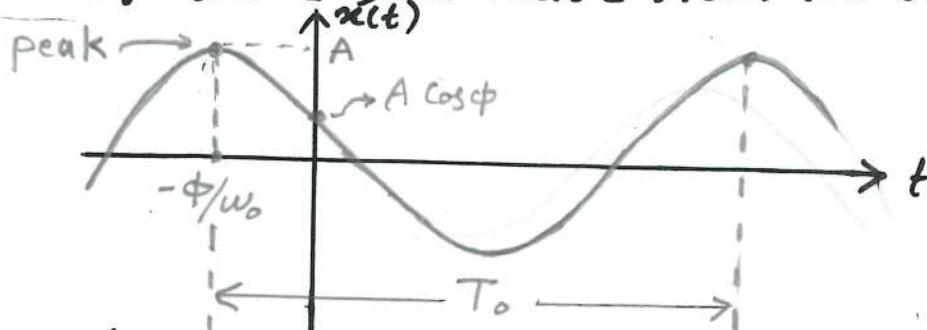
f_0 is the frequency in (Hz). $\boxed{\omega_0 = 2\pi f_0}$

جامعة البصرة - كلية الهندسة
محاضرات
د. قاسم محمد موسى

* The sinusoidal signal repeats itself every time period (T_0) which represents the reciprocal of the frequency:

$$f_0 = \frac{\omega_0}{2\pi} = \frac{1}{T_0} \quad (\text{Hz})$$

* The phase angle represents the shift of the peak value of the cosine wave from the origin.



The peak value of the cosine wave occurs when the angle of the cosine = 0, so:

$$\omega_0 t_{\max} + \phi = 0 \Rightarrow t_{\max} = -\frac{\phi}{\omega_0} \quad (\text{sec.})$$

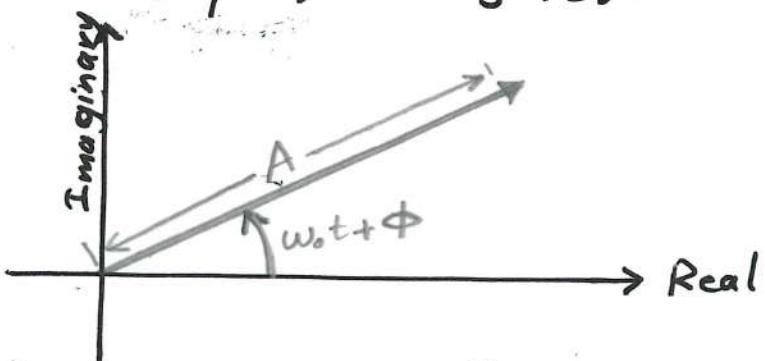
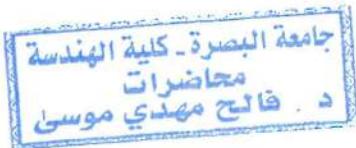
$$\text{OR } \phi = -\omega_0 t_{\max} \quad (\text{rad.})$$

* From Euler's Theorem:

$$e^{j\theta} = \cos\theta + j\sin\theta$$

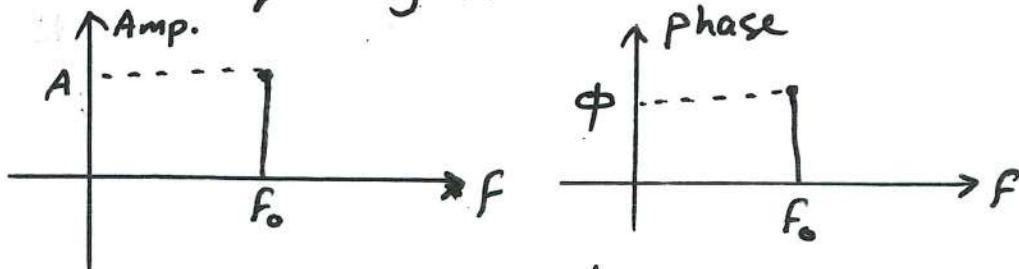
$$A \cos(\omega_0 t + \phi) = A \operatorname{Re}[e^{j(\omega_0 t + \phi)}]$$

- The phasor representation is a time domain plot specified by
 1. Amplitude
 2. phase
 3. rotational frequency



Time-domain phasor of $(A \cos(\omega_0 t + \phi))$

- The line spectrum is a two-plot representation
 1. Amplitude vs Frequency (Amplitude Spectrum)
 2. phase vs frequency (Phase spectrum)



Freq. Domain line spectrum of $(A \cos(\omega_0 t + \phi))$

Notes:

- The reference signal is the cosine wave, so the sine wave should be converted to the cosine using the following identity:
 $\sin \omega t = \cos(\omega t - 90^\circ)$
- Negative amplitude should be converted to $\pm 180^\circ$ phase shift.
 $-A \cos \omega t = A \cos(\omega t \mp 180^\circ)$.

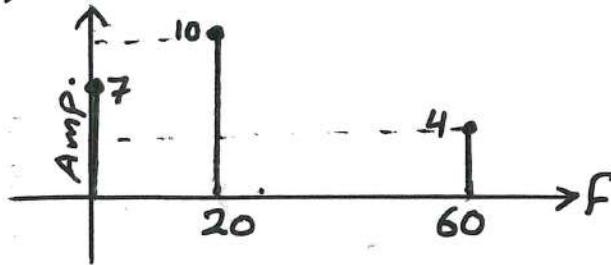
Example: Draw the line spectra of the following wave.

$$x(t) = 7 - 10 \cos(40\pi t - 60^\circ) + 4 \sin 120\pi t$$

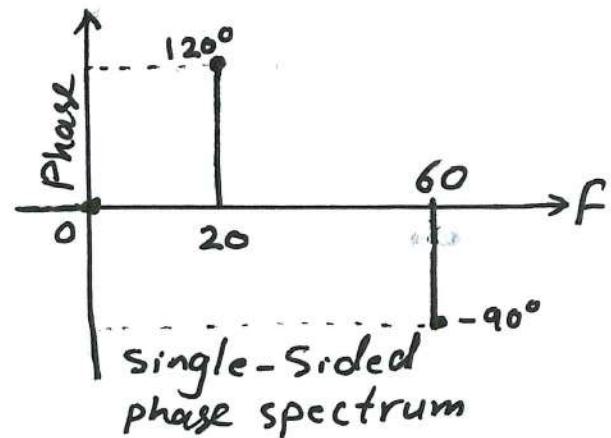
Sol.

$$x(t) = 7 \cos(2\pi 0 t + 0^\circ) + 10 \cos(40\pi t + 120^\circ) + 4 \cos(120\pi t - 90^\circ)$$

$$x(t) = 7 \cos(2\pi \cdot 0t + 0^\circ) + 10 \cos(2\pi \cdot 20t + 120^\circ) + 4 \cos(2\pi \cdot 60t - 90^\circ)$$



Single-sided Amp. Spectrum



Single-Sided phase spectrum

B Double-Sided Spectrum:

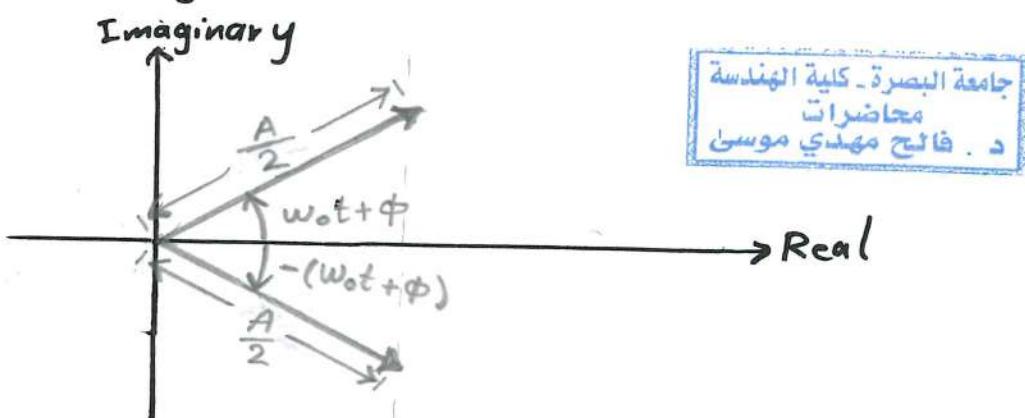
Since $\cos \theta = \frac{1}{2} (e^{j\theta} + e^{-j\theta})$

$$\therefore A \cos(\omega_0 t + \phi) = \frac{A}{2} [e^{j(\omega_0 t + \phi)} + e^{-j(\omega_0 t + \phi)}]$$

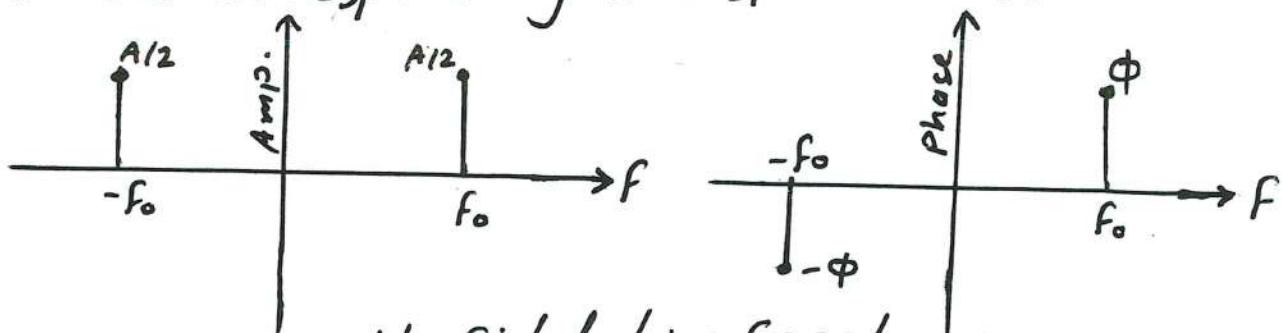
$$A \cos(\omega_0 t + \phi) = \frac{A}{2} [e^{j\omega_0 t} e^{j\phi} + e^{-j\omega_0 t} e^{-j\phi}]$$

$$\text{OR } A \cos(\omega_0 t + \phi) = \frac{A}{2} e^{j\omega_0 t} e^{j\phi} + \frac{A}{2} e^{-j\omega_0 t} e^{-j\phi}$$

The corresponding phasor diagram is :



While the corresponding line spectrum is :



Double-Sided Line Spectrum

* In double-sided spectrum: The Amplitude spectrum has even symmetry, while the phase has odd symmetry.

Example: Draw the double-sided line spectra of the following wave form:

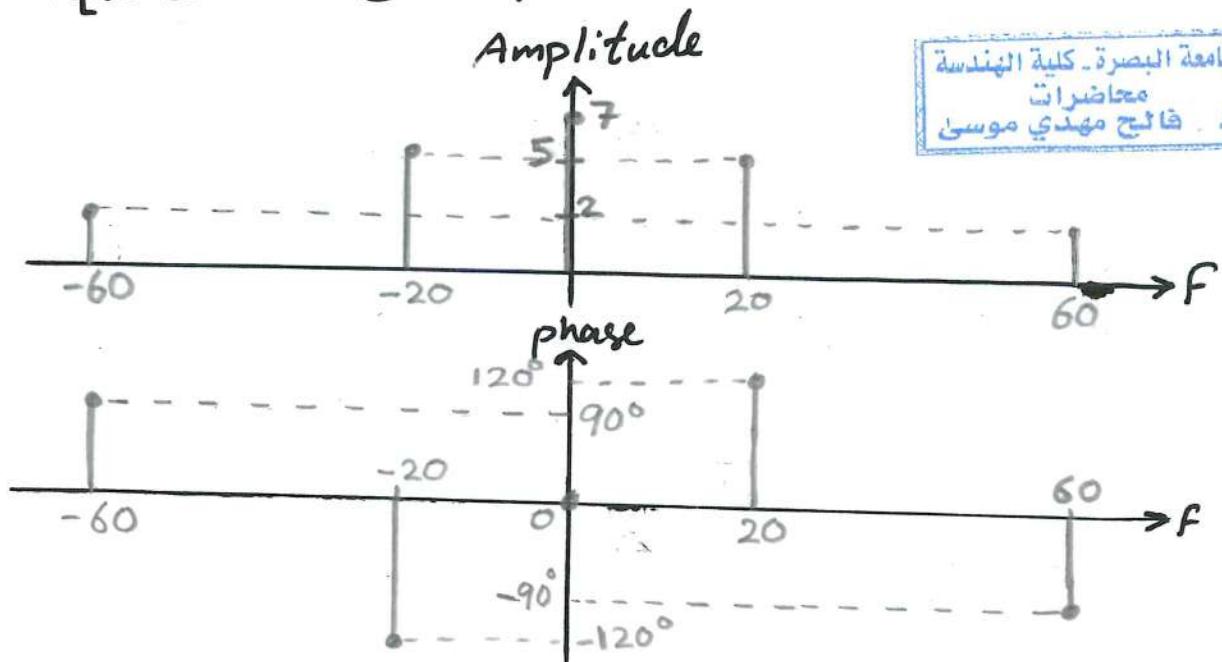
$$x(t) = 7 - 10 \cos(40\pi t - 60^\circ) + 4 \sin 120\pi t$$

Sol.

$$x(t) = 7 \cos(2\pi 0t + 0^\circ) + 10 \cos(2\pi \times 20t + 120^\circ) + 4 \cos(2\pi \times 60t - 90^\circ)$$

$$x(t) = 7 e^{j2\pi 0t} e^{j0^\circ} + \frac{10}{2} [e^{j2\pi \times 20t} e^{j120^\circ} + e^{-j2\pi \times 20t} e^{-j120^\circ}] + \frac{4}{2} [e^{j2\pi \times 60t} e^{-j90^\circ} + e^{-j2\pi \times 60t} e^{+j90^\circ}]$$

$$x(t) = 7 e^{j2\pi 0t} e^{j0^\circ} + [5 e^{j2\pi \times 20t} e^{j120^\circ} + 5 e^{-j2\pi \times 20t} e^{-j120^\circ}] + [2 e^{j2\pi \times 60t} e^{-j90^\circ} + 2 e^{-j2\pi \times 60t} e^{+j90^\circ}]$$



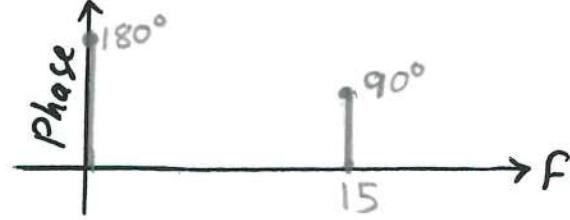
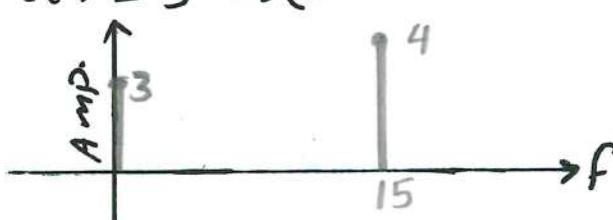
Example: Construct the single-sided and double-sided line spectra of: $x(t) = -3 - 4 \sin 30\pi t$

Sol.

a) Single-Sided line spectrum:

$$x(t) = -3 \cos(2\pi 0t + 0^\circ) - 4 \cos(2\pi \times 15t - 90^\circ)$$

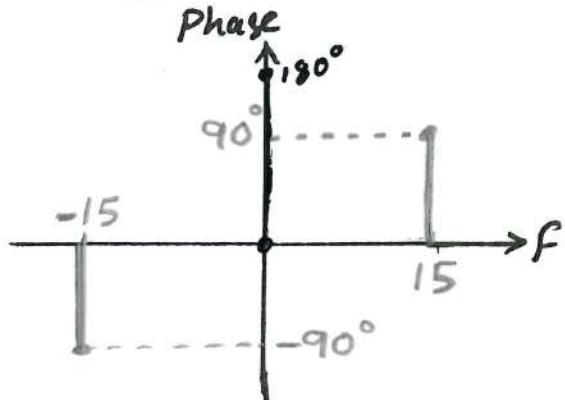
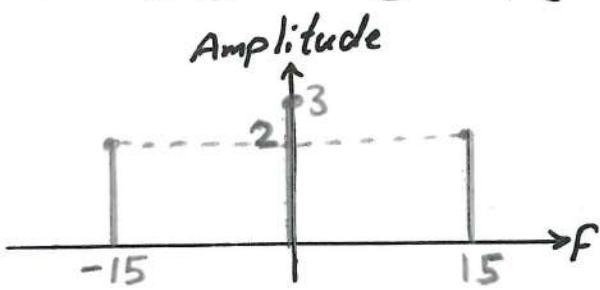
$$x(t) = 3 \cos(2\pi 0t + 180^\circ) + 4 \cos(2\pi \times 15t + 90^\circ)$$



b) Double-Sided Spectrum:

$$x(t) = 3 \cos(2\pi \cdot 0t + 180^\circ) + 4 \cos(2\pi \cdot 15t + 90^\circ)$$

$$x(t) = 3 e^{j2\pi \cdot 0t} e^{j180^\circ} + [2 e^{j2\pi \cdot 15t} e^{j90^\circ} + 2 e^{-j2\pi \cdot 15t} e^{-j90^\circ}]$$



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محاضرات
د. فلاح مهدي موسى

Analysis and Transformation of Signals

11

1. Fourier Series:

- * Any periodic function of time $x(t)$ having a fundamental period (T_0) can be represented by an infinite sum of sinusoidal waveforms.
- * The sinusoidal signal has only a single frequency so it is called "single tone" signal.
- * This means that all the other periodic signals have an infinite frequencies not a single frequency.

Complex (Exponential) Fourier Series:

$$x(t) = D_0 + \sum_{n=-\infty}^{\infty} D_n e^{j2\pi(nf_0)t}$$

where n is integer ($n = \pm 1, \pm 2, \dots$)

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محاضرات
د. فلاح مهدي موسى

$$D_0 = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) dt$$

$$D_n = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-j2\pi(nf_0)t} dt$$

and D_n is complex number $\{D_n = |D_n| e^{j\theta_n}\}$

Notes:

1. IF the periodic signal $x(t)$ is real, then the amplitude spectrum has an even symmetry and the phase spectrum has an odd symmetry.
2. Since the index (n) has only integer values, then the frequency spectra of the periodic signal exist at discrete frequencies (nf_0).

3. D_0 represents the DC component of $x(t)$, and it can be deduced from D_n by setting $n=0$. [2]

Example: Find the complex Fourier series of the following periodic waveform for a) $\tau = \frac{T_0}{2}$. b) $\tau = \frac{T_0}{4}$.

Sol.

$$x(t) = \sum_{n=-\infty}^{\infty} D_n e^{j2\pi(nf_0)t}$$

$$D_n = \frac{1}{T_0} \int_{-T_0/2}^{T_0/2} x(t) e^{-j2\pi(nf_0)t} dt$$

where $f_0 = \frac{1}{T_0}$

$$D_n = \frac{1}{T_0} \int_{-\tau/2}^{\tau/2} 1 \cdot e^{-j2\pi(nf_0)t} dt$$

$$D_n = \frac{1}{T_0} \times \frac{-1}{j2\pi n f_0} e^{-j2\pi(nf_0)t} \Big|_{-\tau/2}^{\tau/2}, \text{ but } f_0 = \frac{1}{T_0}$$

$$\therefore D_n = \frac{-1}{j2\pi n} [e^{-j\pi n f_0 \tau} - e^{+j\pi n f_0 \tau}]$$

$$D_n = \frac{+1}{\pi n} \frac{[e^{j\pi n f_0 \tau} - e^{-j\pi n f_0 \tau}]}{j2}$$

$$D_n = \frac{\sin(\pi n f_0 \tau)}{\pi n} \times \frac{f_0 \tau}{f_0 \tau} = f_0 \tau \frac{\sin(\pi n f_0 \tau)}{\pi n f_0 \tau}$$

$$\therefore D_n = f_0 \tau \operatorname{sinc}(n f_0 \tau)$$

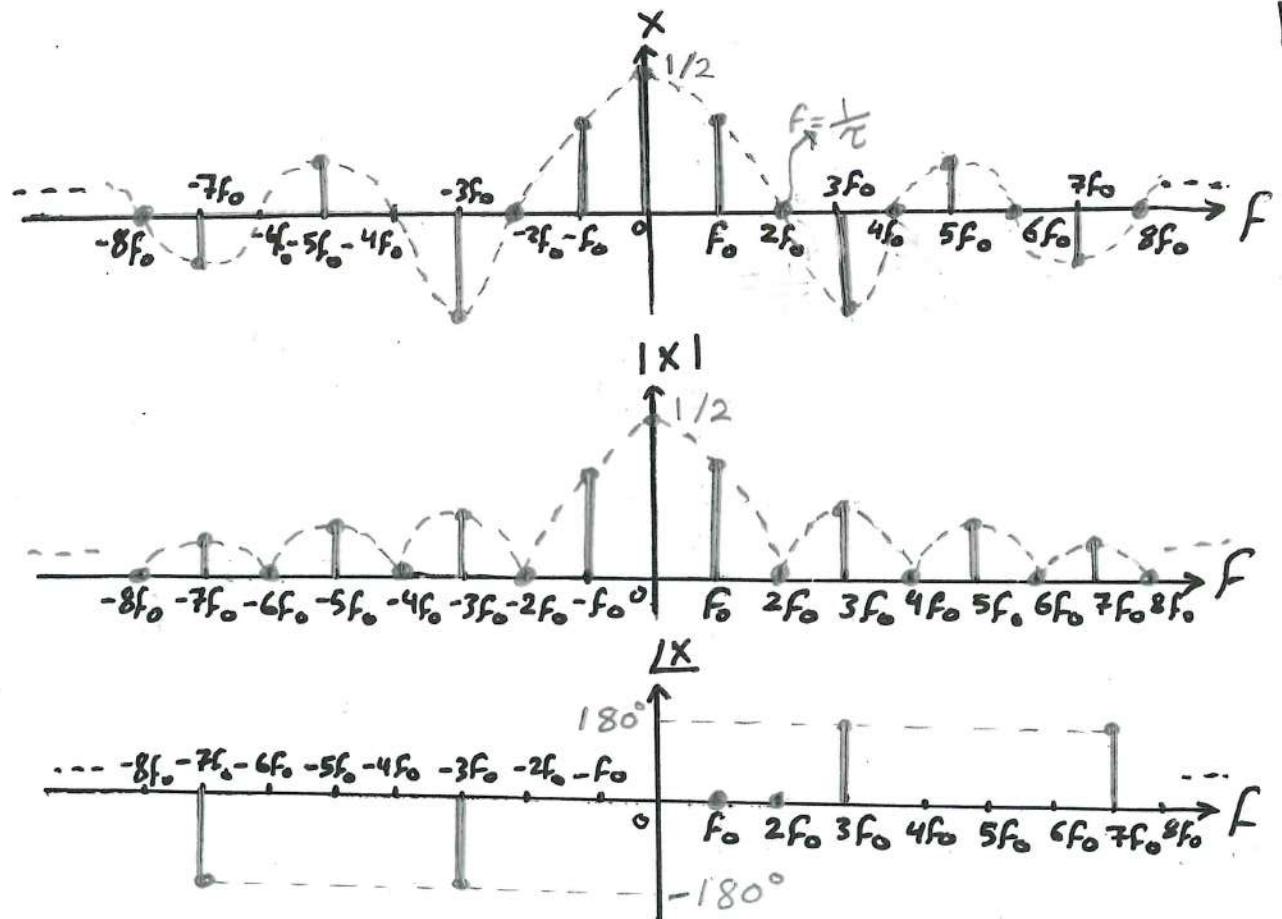
a) for $\tau = \frac{T_0}{2}$:

$$\therefore \frac{\tau}{T_0} = \frac{1}{2} = \tau f_0 \quad \text{where } \frac{\tau}{T_0} \text{ is the duty cycle.}$$

$$\therefore D_n = \frac{1}{2} \operatorname{sinc}\left(\frac{n}{2}\right)$$

$$\therefore x(t) = \sum_{n=-\infty}^{\infty} \left[\frac{1}{2} \operatorname{sinc}\left(\frac{n}{2}\right) \right] e^{j2\pi(nf_0)t}$$

$n = 0, \pm 1, \pm 2, \dots$



b) For $\tau = \frac{T_0}{4}$:

$$\frac{\tau}{T_0} = \frac{1}{4} \Rightarrow \tau f_0 = \frac{1}{4}$$

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محاضرات
د. فائق مهدي موسى

$$D_n = \frac{1}{4} \operatorname{sinc}\left(\frac{n}{4}\right) \Rightarrow x(t) = \sum_{n=-\infty}^{\infty} \left[\frac{1}{4} \operatorname{sinc}\left(\frac{n}{4}\right) \right] e^{j2\pi n f_0 t}$$

